

Higher Order Neural Units for Efficient Adaptive Control of Weakly Nonlinear Systems

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Abstract: The paper reviews the nonlinear polynomial neural architectures (HONUs) and their fundamental supervised batch learning algorithms for both plant identification and neuronal controller training. As a novel contribution to adaptive control with HONUs, Conjugate Gradient batch learning for weakly nonlinear plant identification with HONUs is presented as efficient learning improvement. Further, a straightforward MRAC strategy with efficient controller learning for linear and weakly nonlinear plants is proposed with static HONUs that avoids recurrent computations, and its potentials and limitations with respect to plant nonlinearity are discussed.

Nomenclature

CG ... Conjugate gradient batch learning algorithm
CNU ... Cubic Neural Unit (HONU $r=3$)
colx ... long column vector of polynomial terms
 d ... desired value (setpoint)
LNU ... Linear Neural Unit (HONU $r=1$)
QNU ... Quadratic Neural Unit (HONU $r=2$)
 k ... discrete index of time
L-M ... Levenberg-Marquardt batch learning algorithm
 n, m ... length of vector \mathbf{x}, ξ
 n_y, n_u ... length of recent history of y or u [samples]
 r, γ ... order of polynomial nonlinearity (plant, controller)
 r_o ... control input gain at plant input
 T ... vector transposition
 u ... control input
 \mathbf{w}, \mathbf{v} ... long row vectors of all neural weights (plant, controller)
 \mathbf{x}, ξ ... augmented input vector to HONU (plant, controller)
 \tilde{y} ... neural output from HONU
 y ... controlled output variable (measured)
 y_{ref} ... reference model output
 e ... error between real output and HONU
 e_{ref} ... error between reference model and control loop
 μ ... learning rate

1 INTRODUCTION

Computational intelligence tools such as neural networks or fuzzy systems are booming in the area of system identification and control of dynamical

systems. We can refer to many novel and powerful control approaches, and the most important ones are the Model Predictive Control (MPC), Adaptive Dynamic Programming (ADP) often with extension of reinforcement learning, and Model Reference

Adaptive Control (MRAC). MPC control, e.g. (García et al., 1989; Ławryńczuk, 2009; Morari and H. Lee, 1999) and references therein, depends on a model of controlled system, and the unknown is the actual control input that is being optimized before every controller actuation to achieve the predefined sequence of desired outputs. The core distinction of control via ADP, e.g. (Wang et al., 2009) (WANG et al., 2007) and references therein, is that the controller is designed from heuristically obtained observation of control inputs and corresponding system responses. Then, using the enforced learning with penalization of improper controller action, the controller is directly trained without mathematical analysis of the controlled plant. In principle, ADP is thus suitable also for unstable systems, though the computational costs can be very high. MRAC control, e.g. (Osburn, 1961; Parks, 1966; Narendra and Valavani, 1979; Elbuluk et al., 2002) and references therein, requires model of a system and the unknowns are the parameters (neural weights) of a controller that are trained so the control loop adopts the desired dynamics of a reference model. In this

paper, we focus on MRAC control scheme that is more conservative in sense of adaptive feedback control and generally requires less computing in contrast to ADP. As regards the computational intelligence tools for MRAC, we deal with Higher Order Neural Units (Bukovsky and Homma, 2016; Madan M. Gupta et al., 2013), that represent standalone neural architectures within the same direction of computation that involves polynomial neural networks, e.g. (Nikolaev and Iba, 2006), or higher order neural networks (Ivakhnenko, 1971; Taylor and Coombes, 1993; Kosmatopoulos et al., 1995; Tripathi, 2015). In (Bukovsky et al., 2015) the advantages of HONU feedback controllers are highlighted for real-time application on non-linear fluid based mechanical systems. Further studies focussed on practical applications of HONU-MRAC closed loop control may be recalled in (Benes and Bukovsky, 2014). Except the famous Levenberg-Marquardt algorithm, the Conjugate Gradients (CG) are known to be very fast for linear models and its nonlinear alternatives are studied (Dai and Yuan, 1999; El-Nabarawy et al., 2013; Zhu et al., 2017).

In this paper, we recall Conjugate Gradient learning for HONUs as it can practically accelerate plant identification after first epochs are carried with Levenberg-Marquardt learning. Then, we newly study an approach that use merely static HONUs as a plant models and also as controllers so the recurrent computations are avoided and convergence of controller training is naturally accelerated. Then we newly discuss the drawback of the method that lies in its suitability for weakly nonlinear dynamical systems. The most common symbols and abbreviations are explained in Nomenclature section above, while other terms are explained at their first appearance.

2 PRELIMINARIES ON HONUS

This section recalls HONUs as nonlinear SISO models for such linear or nonlinear dynamical systems, that can be approximated by a HONU with the user-defined polynomial order $r \leq 3$. Thus in this work, a dynamical system is considered weakly nonlinear, if it can be approximated with a recurrent HONU of polynomial order up to $r=3$, i.e. it is a dynamical system, resp. the dynamics in data, that can be reliably captured either by dynamical (recurrent) LNU or QNU or CNU. In other words, a weak nonlinear systems is here such a system that can be approximated via Taylor series expansion of up to 3rd order that corresponds to 3rd order HONU (i.e. CNU). Properly learned dynamics of a plant via HONU is

further necessary to train a linear or nonlinear controller that is realized via another HONU as well. For example, a sinusoidal nonlinearity is an example of strong non-linearity that is difficult or even impossible to fully capture by HONU ($r \leq 3$). For SISO systems, the HONUs of up to third order are defined in a classical form or in a long vector form as defined in Table 1, where the augmented input vector for dynamical system approximation can be defined in (1) and its total length is $n = 1 + n_y + n_u$.

Recall, that y as measured one in (1) implies static a HONU, i.e. static function mapping, because all the values in x are measured ones.

$$\mathbf{x} = \begin{bmatrix} x_0=1 \\ x_1 \\ \vdots \\ x_n \end{bmatrix} = \begin{bmatrix} 1 \\ y(k-1) \\ y(k-2) \\ \vdots \\ y(k-n_y) \\ u(k-1) \\ \vdots \\ u(k-n_u) \end{bmatrix} = \begin{bmatrix} 1 \\ \mathbf{y}_x \\ \mathbf{u}_x \end{bmatrix}, \quad (1)$$

In order to define dynamical (recurrent) HONUs, real y should be replaced with neural outputs, i.e. $y \leftarrow \tilde{y}$ resp. $\mathbf{y}_x \leftarrow \tilde{\mathbf{y}}_x$ in (1), so we would obtain a recurrent neural architecture that can be more difficult to train, but it is necessary for tuning a controller for nonlinear dynamical systems as it is discussed later in this paper. For batch learning for plant identification, simple variant of Levenberg-Marquardt algorithm can be recommended for both static HONUs as well as for recurrent ones so the weight updates $\Delta \mathbf{w}$ can be calculated as follows

$$\Delta \mathbf{w} = (\mathbf{J}^T \cdot \mathbf{J} + \frac{1}{\mu} \cdot \mathbf{I})^{-1} \cdot \mathbf{J}^T \cdot \mathbf{e}, \quad (2)$$

where \mathbf{J} is Jacobian matrix, \mathbf{I} is identity matrix, upper index $^{-1}$ stands for matrix inversion, and the error yields $\mathbf{e} = \mathbf{y} - \tilde{\mathbf{y}}$. Recall that the Jacobian matrix for static HONU, i.e. the input vector (1) with only measured values, is as follows

$$\mathbf{J}(k) = \partial \tilde{y}(k) / \partial \mathbf{w} = \begin{cases} \mathbf{x}^T & \text{for LNU, i.e., } r=1 \\ \mathbf{colx}^T & \text{for QNU, i.e., } r=2 \\ \mathbf{colx}^T & \text{for CNU, i.e., } r=3, \end{cases} \quad (3)$$

where \mathbf{colx} is a long column vector of polynomial terms as indicated in Table 1. Then, equation (3) shows that \mathbf{J} is constant for all training epochs of static HONUs (i.e. when input vector is defined as in

Table 1: Summary of HONUs for weakly nonlinear systems (of up to 3rd polynomial order of nonlinearity).

Order of HONU	HONU (neural output $\tilde{y}(k)$)		details	
	Classical form of HONU =	Long vector form of HONU	$x_0 = 1 \quad \forall r$	
$r = 1$ (LNU)	$\tilde{y} = \sum_{i=0}^n w_i \cdot x_i =$	$= \mathbf{w} \cdot \text{col}^r(\mathbf{x}) = \mathbf{w} \cdot \mathbf{x}$	$\mathbf{w} = [w_0 \ w_1 \ \dots \ w_n]$ $\mathbf{x} = [x_0 \ x_1 \ \dots \ x_n]^T$	
$r = 2$ (QNU)	$\tilde{y} = \sum_{i=0}^n \sum_{j=i}^n w_{i,j} \cdot x_i \cdot x_j =$	$= \mathbf{w} \cdot \text{col}^r(\mathbf{x}) = \mathbf{w} \cdot \text{colx}$	$\text{colx} = [\{x_i x_j\}]$ $\mathbf{w} = [\{w_{i,j}\}]^T$	$i = 0..n$ $j = i..n$
$r = 3$ (CNU)	$\tilde{y} = \sum_{i=0}^n \sum_{j=i}^n \sum_{\kappa=j}^n w_{i,j,\kappa} x_i x_j x_\kappa =$	$= \mathbf{w} \cdot \text{col}^r(\mathbf{x}) = \mathbf{w} \cdot \text{colx}$	$\text{colx} = [\{x_i x_j x_\kappa\}]$ $\mathbf{w} = [\{w_{i,j,\kappa}\}]^T$	$i = 0..n$ $j = i..n$ $\kappa = j..n$

(1)) and that accelerates real time computations (esp. when considering the HONUs are nonlinearly mapping models yet they are linear in parameters). The Jacobian for recurrent HONUs, i.e. when $y \leftarrow \tilde{y}$ in (1), with learning formula (2), and according to per partes derivation rule, generally yields

$$\mathbf{J}(k) = \partial \tilde{y}(k) / \partial \mathbf{w} = (\text{colx}(\tilde{\mathbf{y}}_x))^T + \mathbf{w} \cdot \partial \text{colx}(\tilde{\mathbf{y}}_x) / \partial \mathbf{w}, (4)$$

where the content of a short vector $\tilde{\mathbf{y}}_x$ is apparent from (1). Recurrent HONUs then perform more accurate dynamical approximation of weakly nonlinear dynamical systems from measured data than a static HONU (as it is shown in the experimental section); however, the Jacobian of dynamical HONU varies in time so its rows have to be recalculated at every time sample resulting in varying Jacobian for every training epoch.

Further in this paper, it is shown that static HONU can be sufficient to identify weakly nonlinear dynamical systems so we can use a very efficient, well converging and the most straightforward technique of a feedback controller tuning.

3 CONJUGATE GRADIENT FOR PLANT IDENTIFICATION WITH HONUs

The proper plant identification from measured data is important for further controller tuning. In case of not well conditioned data in Jacobian, the L-M formula (2) needs a small learning rate μ for

convergence and then this gradient method can be slow, i.e. it may need large number of training epochs. Nevertheless, HONUs are linear in parameters and the CG learning can be directly applied to HONUs as it is introduced in this section. Recall, that CG solves the set of equations as follows

$$\mathbf{b} - \mathbf{A} \cdot \mathbf{w} = \mathbf{0}, (5)$$

Where \mathbf{b} is a column vector of constants, \mathbf{A} is positively semi-definite matrix, and \mathbf{w} is a column vector of unknowns (neural weights). Because HONUs are linear in parameters, i.e. the Jacobian (3) is not directly a function of weights, we can restate the training of HONUs as follows

$$\mathbf{y} - \text{colX} \cdot \mathbf{w} = \mathbf{0}, (6)$$

where colX is defined (Bukovsky and Homma, 2016) as follows (assuming all initial conditions are known)

$$\text{colX} = \begin{bmatrix} \text{colx}(k=1)^T \\ \text{colx}(k=2)^T \\ \vdots \\ \text{colx}(k=N)^T \end{bmatrix} = \mathbf{J}, (7)$$

that is in fact the Jacobian of a HONU for all training data (of total length N). Then, by multiplying (6) with colX^T from the left, we obtain

$$\mathbf{b} - \mathbf{A} \cdot \mathbf{w} = \text{colX} \cdot \mathbf{y} - \text{colX}^T \cdot \text{colX} \cdot \mathbf{w} = \mathbf{0}, (8)$$

so we have obtained the positive definite matrix \mathbf{A} and thus we can directly apply the CG learning to both static or recurrent HONUs as follows. For the very first epoch of training we initiate CG with

$$\mathbf{r}_e(\epsilon = 0) = \mathbf{c} - \mathbf{A} \cdot \mathbf{w}(\epsilon = 0), \quad (9)$$

and with

$$\mathbf{p}(\epsilon = 0) = \mathbf{r}_e(\epsilon = 0). \quad (10)$$

Then for further epochs of training, i.e. for $\epsilon > 0$, we calculate parameter

$$\alpha(\epsilon) = \frac{\mathbf{r}_e^T(\epsilon) \cdot \mathbf{r}_e(\epsilon)}{\mathbf{p}^T(\epsilon) \cdot \mathbf{A} \cdot \mathbf{p}(\epsilon)}, \quad (11)$$

and then we can update the weights as

$$\mathbf{w}(\epsilon + 1) = \mathbf{w}(\epsilon) + \alpha(\epsilon) \cdot \mathbf{p}(\epsilon), \quad (12)$$

where $\Delta \mathbf{w} = \alpha(\epsilon) \cdot \mathbf{p}(\epsilon)$ and other CG parameters for next training epoch are then calculated or updated as follows

$$\mathbf{r}_e(\epsilon + 1) = \mathbf{r}_e(\epsilon) - \alpha(\epsilon) \cdot \mathbf{A} \cdot \mathbf{p}(\epsilon), \quad (13)$$

and similarly to Fletcher-Reeves nonlinear CG method

$$\beta(\epsilon) = \frac{\mathbf{r}_e^T(\epsilon + 1) \cdot \mathbf{r}_e(\epsilon + 1)}{\mathbf{r}_e^T(\epsilon) \cdot \mathbf{r}_e(\epsilon)}, \quad (14)$$

and finally

$$\mathbf{p}(\epsilon + 1) = \mathbf{r}_e(\epsilon + 1) + \beta(\epsilon) \cdot \mathbf{p}(\epsilon). \quad (15)$$

This section extended the classical Conjugate Gradient learning (as known for linear systems) to HONUs due to their in parameter linearity. This batch learning algorithm (9)-(15) has no adjustable parameters except the initial neural weights and the number of learning epochs; furthermore, the training of both static or dynamical HONU can be rapidly accelerated in comparison to L-M algorithm (2). As demonstrated in experimental section in this paper, it is suggested to use L-M learning for a few initial training epochs and then switch for CG to rapidly accelerate plant identification. Notice that in a closed loop, the presented CG for HONU is not so suitable for controller weights \mathbf{v} training as the symmetric positive definite matrix is not so achievable and CG for a controller then becomes much more complicated task. Thus L-M and CG are both applicable to plant identification via static or dynamic HONUs, while only L-M is the most comprehensible training algorithm for HONUs as controllers within the MRAC control scheme (Figure 2))

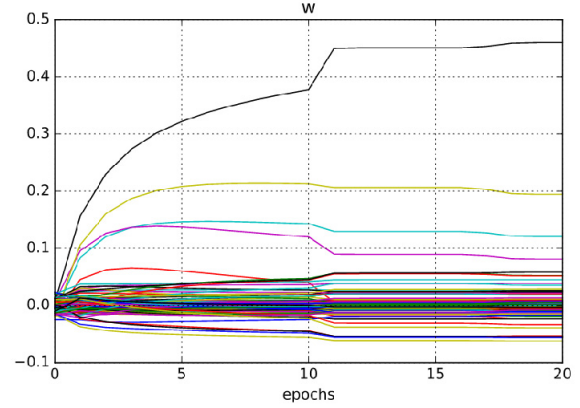


Figure 1: Training of static CNU for system in Figure 3 starting with L-M learning accelerated with CG learning for training epochs ≥ 10 .

4 BATCH LEARNING STRATEGY FOR CONTROL WITH STATIC HONU AS A PLANT MODEL

The most typical control loop for SISO dynamical systems with HONUs can be according to Figure2, where we assume normalized (z-scored) magnitudes

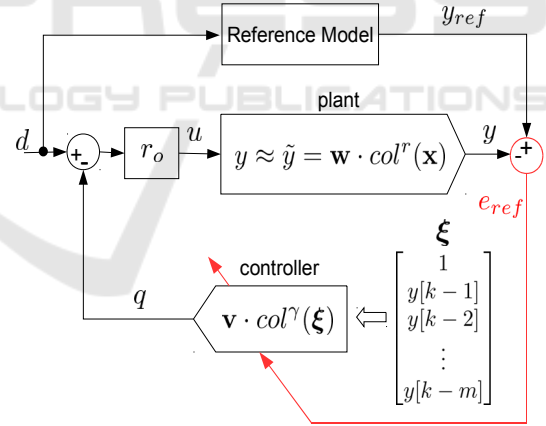


Figure 2: Discrete time Model Reference Adaptive Control (MRAC) loop with two HONUs: as plant model and as a nonlinear state feedback controller.

of input and output variables, i.e. of d , y , and y_{ref} and where the input gain r_o can be also adaptive and it compensates for the true static gain of the controlled plant. In our previous works on HONUs with the MRAC control scheme in Figure 2, we have usually identified plants with recurrent HONUs and then we used a trained HONU as a dynamical plant model for further controller-tuning computations, i.e.,

including for the control loop simulation with HONU as a dynamical model of a plant. It means that for the control law

$$u(k) = r_o \cdot (d(k) - q(k)), \quad (16)$$

the controller weights are optimized (trained) via L-M learning rule as follows

$$\Delta \mathbf{v} = (\mathbf{J}_v^T \cdot \mathbf{J}_v + \frac{1}{\mu_v} \cdot \mathbf{I}_v)^{-1} \cdot \mathbf{J}_v^T \cdot \mathbf{e}_{ref}, \quad (17)$$

where the subscript v indicates it is the controller learning rule, and where the Jacobian matrix \mathbf{J}_v involves recurrent backpropagation computations of partial derivatives (with properly applied time indexes as follows

$$\mathbf{J}_v[k, :] = \frac{\partial \tilde{y}(k)}{\partial \mathbf{v}} = \frac{\partial \tilde{y}}{\partial u} \frac{\partial u}{\partial q} \frac{\partial q}{\partial \mathbf{v}}, \quad (18)$$

so we should calculate the Jacobian (18) recurrently because

$$\frac{\partial q(k)}{\partial \mathbf{v}} = (\text{col}^\gamma(\boldsymbol{\xi}))^T + \mathbf{v} \cdot \begin{bmatrix} \mathbf{0} \\ \frac{\partial \tilde{y}(k-1)}{\partial \mathbf{v}} \\ \frac{\partial \tilde{y}(k-2)}{\partial \mathbf{v}} \\ \vdots \\ \frac{\partial \tilde{y}(k-m)}{\partial \mathbf{v}} \end{bmatrix}, \quad (19)$$

where the utmost right matrix has its dimension of $(1+m) \times n_v$ and n_v denotes the total number of neural controller weights (i.e. the length of \mathbf{v}). For the above controller training (16)-(19) (with dynamical HONU as plant model), both the input vectors \mathbf{x} in (1) and $\boldsymbol{\xi}$ in (Figure 2). Based on our contemporary knowledge (Benes and Bukovsky, 2014; Benes et al., 2014; Bukovsky et al., 2015, n.d.), the above control scheme with dynamic plant HONU works very well if the dynamical HONU can be trained to identify the plant, i.e. if the plant is not more nonlinear than as it could be captured by the dynamical HONU. In other words, we always suggest to try to identify the plant with a dynamical HONU for control tuning for MRAC approach. Nevertheless, in case of weakly nonlinear plants, i.e. plants that can be well identified by the dynamical HONUs due to not too high degree of nonlinearity, we may avoid the recurrent computations and we can train the controller weights in much more straightforward and computationally efficient strategy as follows.

At the beginning of controller training and before the plant identification, we always need a reasonable training data, here we define the desired variable d that we can feed into the plant to measure the corresponding plant output data y , and we also simulate the reference model output y_{ref} . Recall that the objective is to modify the control loop so it adopts the reference model behavior once the controller is properly trained. In other words, we desire that the trained controller outputs a proper value q so in the next corresponding future steps, the controlled plant y matches the reference model output y_{ref} .

$$\mathbf{x} = \begin{bmatrix} 1 \\ y_{ref}[k-1] \\ y_{ref}[k-2] \\ \vdots \\ y_{ref}[k-n_y] \\ u[k-1] \\ \vdots \\ u[k-n_u] \end{bmatrix}, \quad \boldsymbol{\xi} = \begin{bmatrix} 1 \\ y_{ref}(k-1) \\ y_{ref}(k-2) \\ \vdots \\ y_{ref}(k-m) \end{bmatrix}, \quad (20)$$

Thus, if we feed the reference values directly into the input vectors of plant (1) and controller (Figure2) as in (20) i.e. we do not need to simulate the closed loop output with a recurrent HONU model of a plant but instead we use directly the apriori given reference model values, then we can update the controller weights directly and the k^{th} row of Jacobian (18) can be directly evaluated in (21) where the weights \mathbf{w} were obtained by L-M and/or CG learning algorithm with static HONU as a plant model. Thus, we have avoided the recurrent computations in both plant identification as well as controller training and the computational efficiency and convergence of the learning is boosted.

$$\frac{\partial \tilde{y}(k)}{\partial u} = \mathbf{w} \cdot \begin{bmatrix} \mathbf{0} \\ \vdots \\ \mathbf{0} \\ \frac{\partial u(k-1)}{\partial \mathbf{v}} \\ \vdots \\ \frac{\partial u(k-n_u)}{\partial \mathbf{v}} \end{bmatrix}; \text{ where} \quad (21)$$

$$\frac{\partial u(k-1)}{\partial \mathbf{v}} = -r_o \cdot (\text{col}^\gamma(\boldsymbol{\xi}))^T$$

However as mentioned above, the price for this boosted controller design is that this way trained

controllers are limited to linear or weakly nonlinear plants, i.e. to such plants for which the static HONU sufficiently approximates the system from data (otherwise recurrent HONU has to be used for plant identification for the control scheme and it can be more difficult to train dynamical HONU properly or even impossible depending on nonlinearity of the approximated system).

5 EXPERIMENTAL ANALYSIS

The plant dynamics was first identified from input-output data with static HONU and another static HONU(s) were trained as controllers (Figure 2) with L-M training, so the recurrent computations and plant simulation during the plant identification and controller training were avoided. All the computations were implemented in Python (2.7) including the continuous time simulation of plants and closed control loops (Scipy odeint with default setups unless noted otherwise).

5.1 Linear Oscillating Dynamical System

To demonstrate the efficiency of the proposed control approach with HONU for oscillatory linear dynamical systems, a SISO plant defined with the Laplace transfer function is chosen as follows

$$G(s) = \frac{50s^2 + 10s + 5E4}{50s^4 + 500s^3 + 5E4s^2 + 4E4s + 2E6}, \quad (22)$$

where s is the Laplace operator. The training data were simulated with sampling $\Delta t = 1$ [time unit] and its dynamics was approximated with a single static LNU (i.e. HONU $r=1$, $n_y=5$, $n_u=5$, $\mu=0.1$, 300 training epochs of L-M training) for the same constant sampling. To enhance the controller performance for this dynamical system with complex conjugate poles and zeros, two parallel LNUs were used as a controller to enhanced control performance as introduced recently in (Bukovsky et al., n.d.). Recall that in such case, the controller consisted of two summed LNUs so the control law (16) yielded

$$u(k) = r_o \cdot (d(k) - q_1(k) - q_2(k)), \quad (23)$$

where q_1 and q_2 are outputs of two parallel LNUs (while only a single one is shown in Figure 2). Both controller LNUs were trained with L-M learning

with input vector setups $m_1=5$, $m_2=5$, and with the same learning rate $\mu_v = 1E4$ for the weights and for r_o within 30 training epochs. The performance of the trained controller is shown in Figure 4.

5.2 Weakly Nonlinear System

As an example of a weakly nonlinear system, the second-order dynamics plant is chosen as defined in Figure 3,

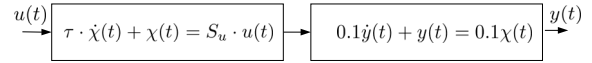


Figure 3: A weakly nonlinear plant where $\tau = \tau(u)$ and $S_u = S_u(y)$ are nonlinear functions (24).

where the time constant τ and the static gain S_u of the first subsystem are nonlinear functions as follows

$$\tau = \tau(u) = 0.2 + |u| \cdot 0.1, \quad S_u = S_u(y) = 0.7 + |y| \cdot 0.2. \quad (24)$$

The continuous-time plant (to obtain “measured” data) and the continuous control loop were simulated with sampling $\Delta t = 0.01$ [time unit] with forward Euler method. The plant was identified from “measured” data with static CNU (HONU $r=3$) with setups $n_y=3$, $n_u=3$ and with initial 10 epochs of L-M with $\mu=0.01$ followed with 10 epochs of CG for accelerated learning (see the accelerated convergence in Figure 1). Further a single CNU as a controller (as in Figure 2) with setup $m=2$ and also r_o were trained by L-M (10 epochs, $\mu_v=1E6$). The sampling of plant identification and controller actuation was here $\Delta t_{HONU}=0.5$ [time unit]. The performance of the trained control loop compared to the plant without a controller is shown in Figure 5.

6 DISCUSSION

To discuss the limitations of HONUs of up to 3rd polynomial order, i.e. $r \leq 3$, as we have investigated them for nonlinear dynamical systems up to now, let’s discuss identification and control of an inverted pendulum model (25) with HONUs in this section. The model (23) is adopted from (Liu and Wei, 2014); however, it is modified with increased friction to at least stabilize the system first to obtain training data for plant identification. It should be highlighted, that the MRAC control approach depends on plant model, here learned from training

data, and such data can not be usually obtained from unstable systems. The modified (stabilized) discrete time inverted pendulum model is then as follows

$$\begin{aligned}\chi_1(k+1) &= \chi_1(k) + 0.1 \cdot \chi_2(k) \\ \chi_2(k+1) &= -0.49 \cdot \sin(\chi_1(k)) + \\ &+ (1 - 0.1 f_d) \cdot \chi_2(k) + 0.1 \cdot u(k)\end{aligned}\quad (25)$$

where the measured output is simulated as $y = \chi_1$, $f_d = 0.6$ is the friction and u is control input. A successful setup for at least imperfect control of this plant (see Figure 6) was found at doubled sampling with static CNU for identification ($r=3$, $n_u=n_i=4$, trained with 20 epochs of L-M ($\mu=1$) followed with 20 epochs of CG) and two parallel static HONUs (LNU and CNU) as a controller, as discussed in subsection 5.1 (23), both with $m=4$ and with 30 epochs of L-M training ($\mu_v=1$). The successful setup for control of this more strongly nonlinear plant was not so trivial to find as it was for the previous two weakly nonlinear plants. The control result for (25) is shown in Figure 6.

7 CONCLUSIONS

In this paper, we have presented MRAC control strategy with purely static HONUs that avoids recurrent computations and thus improves convergence of controller training. As an aside, the Conjugate Gradient was presented for HONUs as it can accelerate plant identification with HONUs. This adaptive control technique is easy to implement and it was shown working for weakly nonlinear dynamical systems, i.e. such systems that can be well approximated with HONUs of appropriate polynomial order (here < 3). Investigating this straight control approach with HONUs for strongly nonlinear systems is still a challenge.

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APPENDIX

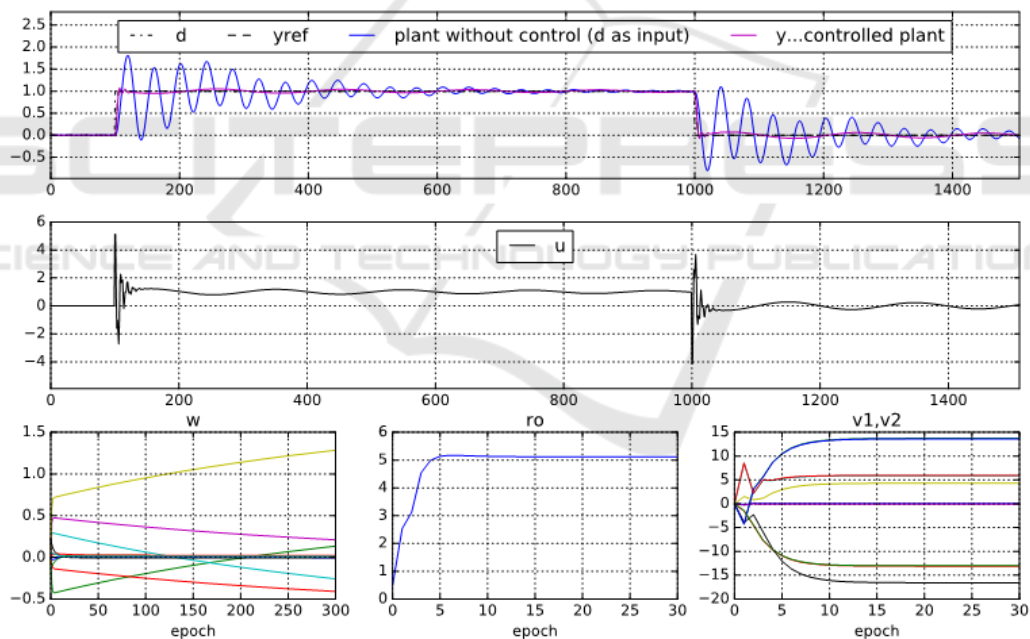


Figure 4: Results for linear oscillatory plant (22) via static LNU as a plant model (trained via L-M) and two parallel LNUs as a feedback controllers (trained via L-M); the control loop output follows the desired unit steps reference signal (y_{ref}) (the upper plot), the control input (2nd from top), bottom axes shows training by L-M batch learning of static LNU for plant (bottom left), input gain (middle), and static LNU as controller (bottom right).

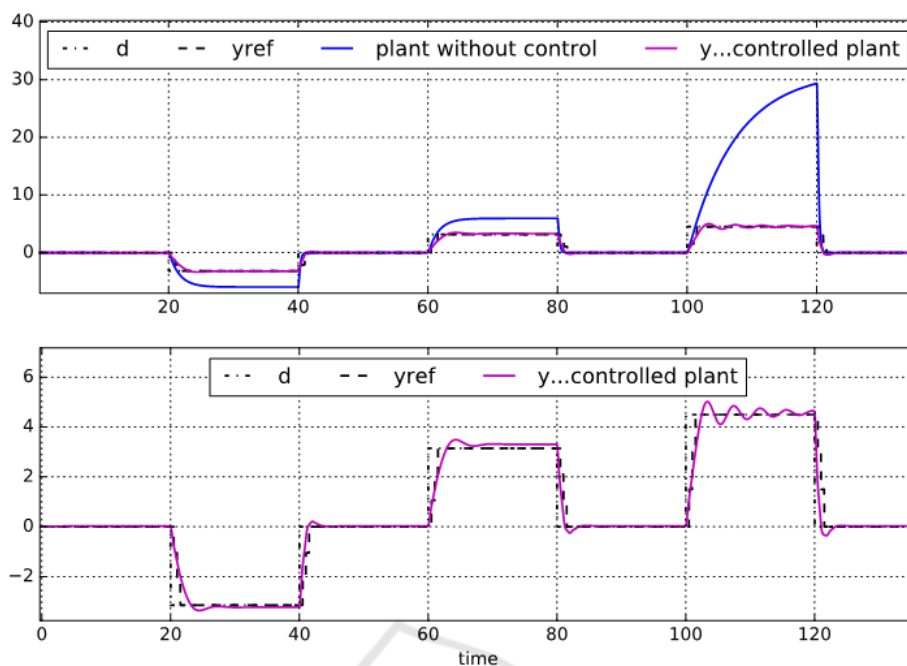


Figure 5: Results for control of weakly nonlinear dynamical system (Subsection 5.2) with variable time constant and static gain (24), static CNU as a plant model (trained via L-M + CG, Figure 1) and static CNU controller (via L-M) (the bottom axes show the detail of the desired behavior with the trained control loop output).

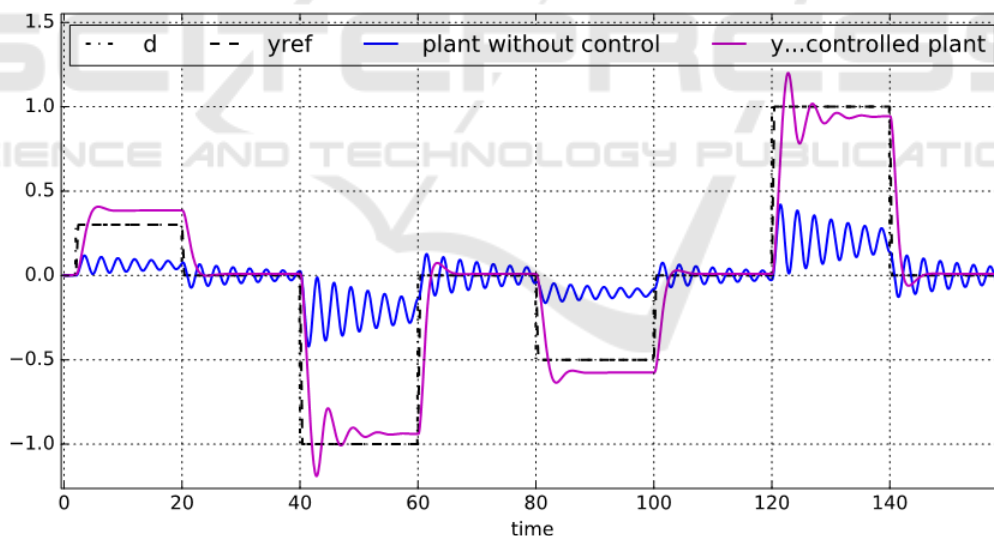


Figure 6: Control of plant (25); with the increasing desired value (dashed), the accurate control is more difficult to achieve as the sinusoidal nonlinearity becomes stronger with increasing magnitude of desired value, so it demonstrates current limits and challenges of the identification and control of strongly nonlinear systems with HONUs.