## ON SOME PECULIARITIES OF CLUSTER ANALYSIS OF PERIODIC SIGNALS

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Abstract: An important task of geophysical research is in the answer to the question about the quality of signals, i.e., estimating the locus of the signal and the degree of their presence in noises. Such indications determine the degree of trust to consequent estimations (e.g., estimations of wave arrival times). As seismic data are periodic signals in their nature, conventional means for examining such signals are Fourier and spectral analyses. However, this method does not allow us to clear up questions about probability of signals presence and their locus in the recorded data. We consider another approach – the cluster analysis of periodic signals, propose the formal conditions which must be satisfied by a period of signal existence, and give some results of analysis of real data recorded in field conditions.

## **1 INTRODUCTION**

The prompting motive of our research is the needs in noisy geophysical data analysis. The basis of such data is periodic (harmonic or frequency-modulated) signals recorded at discrete instants of time. However, the corresponding signals are widely used, and an appropriate research can be of interest in other fields of activity.

In practice, geophysical data are recorded in field conditions. The point is that in the process of wave propagation and in recording data one or another type of errors takes place. Therefore, analysis of such data demands a special attention. Usually, in this case researchers attract a harmonic (I. I. Gurvich, and G. N. Boganic, 1980) or a spectral analysis (E. A. Davidova, and others, 2002). However, an appropriate approach cannot decide the dilemma "time-frequency" (the spectrum components are listed in a domain, where the time scale is absent).

Currently, methods of wavelet analysis and transformation (A. A. Nikitin, 2006, E. Baziw, 1994) are of interest to researchers. Here, time localization of the signal frequency components can be found. Essentially, such an approach is an analog to convolution or linear concordant filtration, or, in other words, it is a development of the window Fourier analysis.

We will treat periodic signals as time series and consider another approach, based on the cluster analysis. To the best of our knowledge, the notion of a cluster is used by few of authors to analyze periodic signals (Znak V. I. and Grachev O. V., 2009).

## 2 METHODOLOGY OF CLUSTER ANALYSIS OF PERIODIC SIGNALS

We can treat the time evolvent of a periodic signal on a plane as a specific image, and set a problem of studying some or other its features. However, such an image becomes considerably complicated in the presence of noises.

The problem can be simplified if an image of some integrated estimation of a signal is used as an object of analysis. Here, we offer to employ an estimation of a standard deviation (dispersion) on some running basis. The behaviour of such estimations as time function  $\sigma(t)$  will reflect the energy distribution of a signal in the region of their existence. Then, the evolvent of function  $\sigma(t)$  on a corresponding 2-D plane can be considered as an image of a cluster formation. Features of such an image are of interest for the purposes of analysis of periodic signals.

226 Znak V. (2010). ON SOME PECULIARITIES OF CLUSTER ANALYSIS OF PERIODIC SIGNALS. In *Proceedings of the 7th International Conference on Informatics in Control, Automation and Robotics*, pages 226-229 DOI: 10.5220/0002911502260229 Copyright © SciTePress As the above estimation we use

$$\sigma(t) = \sigma_k(L) = \sqrt{\sum_{j \in L} (x_j - M(L))^2} / L$$
 (1)

where *L* is odd, *x* is signal values, j=(L-1)/2,..., N-(L-1)/2 (*N* is a signal length). We will assume  $\sigma(t)$  is integers.

Let  $\hat{\sigma}$  be the uppermost dispersion value:  $0 \le \sigma_k \le \hat{\sigma}$  and  $t_k$  is an instant of time. Then, some integer  $h \ (0 \le h \le \hat{\sigma})$  will be called a "threshold". Thus, we juxtaposed with our signal estimations a grid  $h_l \times \sigma_k$  on a 2-D plane, which will be denoted as  $Q: h=0,..., \hat{\sigma}$ ; k=(L-1)/2,..., N-(L-1)/2). Further, we will suppose that each point of the grid represents an event  $q_k(h) \in Q$ , where  $q \in (0,1)$ :

$$q_{k}(h) \begin{cases} 1, \text{ if } \sigma_{k} \ge h_{l}, \\ 0, \text{ if } \sigma_{k} < h_{l}. \end{cases}$$
(2)

For any threshold *h*, the respective subset  $Q_r(h) \subset Q$  for the adjacent instants will be called a cluster if for all the events  $q_k$  of  $Q_r(h)$  the corresponding  $\sigma_k$  is greater or equal to the *h*, i.e.:  $\forall q \in Q_r(h)$ : q=1. The cardinal number of such cluster is

$$\beta_r(h) = \sum_{q \in Q_r(h)} q, r = 1,..., m(h)$$
 (3)

and locus in time is  $\Delta t_r(h) = t_{1(r)}(h) \div t_{n(r)}(h)$ .

Naturally, both the quantity of such clusters and the cardinal number of each cluster depend on the threshold value.

We can speak about two clusters of the two neighboring thresholds that a cluster  $Q_{s(r)}(h+1)$  is a child of  $Q_r(h)$  if they are intersecting in time:  $\Delta t_r(h) \& \Delta t_{s(r)}(h+1) \neq 0$ . We will pool such clusters and call them a cluster family. The cardinal number of this cluster family is

$$b_r(h) = \beta_r(h) + \sum_{s(r)=1}^{n(r)} \beta_{s(r)}(h+1)$$
(4)

etc. Let

$$B(h) = \sum_{r=1}^{m(h)} b_r(h)$$
 (5)

be a common cardinal number of cluster families on the threshold h. Then the relation

$$P_r(h) = b_r(h) / B(h) \tag{6}$$

will be called a representative probability of the family  $Q_r(h)$ .

Let us consider a series of functions  $P_r(h)$ , r=1,..., m(h),  $h=1,..., \hat{\sigma}$ . We expect that the behavior of such functions reflects the degree of the presence of a signal in noise. At the same time, they are tied to subjects, which have their own locus in time.

The matter of the problem is to investigate the behaviour of these functions for answering the questions about the degree of the presence of a periodic signal in noise, and its locus in noisy data.

## 3 ON STUDING THE SIGNAL EXISTENCE

Let us consider some cluster formation  $\sigma_k(L)$  as an image  $\aleph$  under condition of any running basis *L* (Fig. 1).

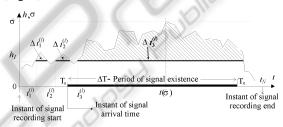


Figure 1: An example of the mapping of the dispersion estimations with a running basis.

Now, turn to the question about picking out the most informative threshold with regard to a representative probabilities. For answer on the question, we will study situations beginning from threshold h=0. However, first of all, we will make some remarks based on the nature of a signal in question. We suppose:

1) A process of signal recording in time (t=0) begins before a signal arrival, i.e.,  $T_0>0$ , where  $T_0$  is an instant of a signal arrival time.

2) A signal on the dispersion estimator input is  $y=s+\xi$ , where *s* is a source signal, and  $\xi$  is an additive white noise with zero mean Gaussian distribution. Let  $t_N$  be a signal recording period, and  $\Delta T$  a signal existence period. Then, the following conclusion is a consequence of such supposition: the probability of localizing the uppermost dispersion value  $k(\hat{\sigma})$  on  $\Delta T$  is proportional to the ratios  $t_N/\Delta T$  and to the signal-to-noise ratio, i.e.,  $P(\hat{\sigma}(k) \in \Delta T) \sim t_N/\Delta T \& s/\xi = f(t_N/\Delta T, s/\xi)$ . (a more exact dependence needs a separate attention).

Now, we will study a cluster families beginning with threshold  $h_0=0$ . We can say, that the threshold  $h_0=0$  is non-informative for us because we have  $t_1(0)=0$  for a single clusters family  $Q_1(h_0)$ (obviously,  $P_1(h_0)=1.0$ ). We can say the same with regard to the threshold  $h_1=1$  if the same conditions  $t_1(h_1)=0$  for single  $Q_1(h_1)$  ( $P_1(h_1)=1$ ) are fulfilled, and so on.

Let, for the first time in threshold raising,  $h_l$  be such a threshold, where  $t_1(h_l) > 0$  for  $m(h_l) \ge 1$ . In this case, we will have the three sets: 1) a set of instants of time of the beginning of cluster families  $t_{1(r)}(h_l)$ , 2) a set of periods of existence of the appropriate cluster families  $\Delta t_{(r)}(h_l)$ , and, 3) representative probabilities of the appropriate families  $P_r(h_l)$  ( $r=1,...,m(h_l)$ ).

Here, the following conditions are fulfilled:

- i) one (or more) of such instant of time at which the condition  $t_{1(r)}(h_l) > 0$  is fulfilled;
- ii) a set of probabilities includes such P<sub>r</sub>(h<sub>l</sub>), that the condition P<sub>r</sub>(h<sub>l</sub>)=max is fulfilled;
- iii) a set of periods of existence of cluster families includes such  $\Delta t_r(h_l)$  that the condition  $t(\hat{\sigma}) \in \Delta t_r(h_l)$  is fulfilled  $(1 \le r \le m(h_l))$ .

We will suppose that a locus of signal existence is reflected by such a period of existence of the cluster family  $\Delta t_r(h_l)$ , which fulfils (obeys) the following conditions:

$$t_{1(r)}(h_l) > 0, P_r(h_l) = \max, t(\hat{\sigma}) \in \Delta t_r(h_l),$$

(7)

where  $\Delta t_r(h_l)$  is  $t_{1(r)}(h_l) \div t_{n(r)}(h_l)$ .

# 4 AN EXAMPLE OF STUDING A SIGNAL

The methodology in question was used for analysis of data recorded in the course of monitoring (in 2007) of the Karabetov mud volcano on Mt. of Taman Province (data recorded by Z-component of receivers for profile line T1, results of field experiments are currently accessible on the web\_site http://opg.sscc.ru). Seismic (or vibro-) records were recorded from 10 T vibratory source with a frequency band of 10–64 Hz, and with a sampling frequency = 0.004 sec. Appropriate data can be found at http://opg.sscc.ru/db.

The results obtained are given in the Table 1. The columns of this Table include distances between a vibratory source and a receiver (S),  $h_{\min} \div h_{\max} = \hat{\sigma}$ , periods of existence of the appropriate cluster families  $t_1 \div t_2$ , and appropriate estimations of representative probabilities (P) for running basis

Table 1: Estimations of signal locus in time.

S n	n. L	$h_{\min} \div h_{\max}$	$t_1 \div t_2$	Р
2363	25	6÷415	3933÷19988	0.94
	75	14÷396	3689÷15669	0.89
	125	16÷390	1520÷14247	0.91
2415	25	9÷305	2041÷19502	0.95
	75	16÷292	2284÷19672	0.96
	125	28÷279	3901÷9316	0.81
2461	25	13÷156	3173÷9917	0.45
	75	24÷133	3190÷10305	0.50
	125	27÷121	3210÷10471	0.52
2557	25	3÷64	3315÷19988	0.90
	75	6÷53	3481÷19963	0.92
	125	6÷46	3455÷19938	0.92
2601	25	5÷165	4045÷19988	0.88
	75	11÷151	4060÷18122	0.85
	125	15÷132	3741÷18114	0.88
2647	25	8÷98	3493÷8855	0.42
	75	12÷88	3593÷14414	0.73
	125	14÷86	2213÷13365	0.77
2698	25	2÷13	8642÷19988	0.73
	75	3÷13	8832÷19963	0,77
	125	3÷13	8807÷19938	0.77
2749	25	6÷87	1410÷12065	0.59
°.1	75	8÷84	1150÷19963	0.98
1.4	125	9÷82	1170÷19938	0.99
2796	25	6÷87	2637÷15919	0.89
	75	10÷78	972÷15800	0.98
	125	10÷71	360÷16047	0.99
2845	25	15÷111	3508÷10413	0.58
	75	21÷105	569÷11594	0.82
	125	22÷100	898÷11615	0.83
2894	25	6÷51	3897÷18815	0.89
	75	9÷46	3889÷15243	0.88
	125	10÷43	3863÷11205	0.59
2999	25	5÷76	9532÷19988	0.51
	75	8÷67	9573÷19963	0.5
	125	8÷56	1824÷19938	0.96
3046	25	6÷60	5646÷19988	0.80
	75	8÷52	2209÷19963	0.97
	125	8÷47	2185÷19938	0.97
3095	25	5÷111	3987÷19988	0.93
	75	7÷90	3988÷19963	0.95
	125	7÷76	3963÷19938	0.95
3141	25	12÷67	3665÷11019	0.65
	75	17÷61	3838÷9707	0.71
	125	18÷54	4054÷9388	0.74
3198	25	12÷74	4945÷19988	0.81
	75	`8÷66	6011÷9346	0.43
	125	18÷64	4288÷19938	0.87

#### $L \in \{25, 75, 125\}.$

Estimations of signal locus in time for  $S\sim0$  m are given in the Table 2 (all the representative probabilities are equal to 1).

The processing and analysis data were obtained by means of interactive computer system of designing and support of one-dimensional weighed order statistics filters (V. I. Znak, 2009).

Table 2: Estimations of signal locus in time for S~0 m.

L	$h_{min} \div h_{max}$	$t_1 \div t_2$
25	6÷415	3870÷18883
75	14÷396	3846÷18908
125	16÷390	3821÷18933

By way of example, an image of investigated signal (for S=2647 m) and appropriate dispersions is shown in Fig. 2.

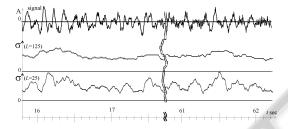


Figure 2: An image of signal and appropriate dispersions for L=25, L=125 (S=2647 m).

Estimations of the data from the Table 1 are given in Fig. 3.

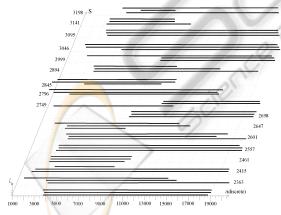


Figure 3: Estimations of the time locus of the signal for running basis  $L \in \{25, 75, 125\}$  and for different distances.

### **5** CONCLUSIONS

We have considered the approach of cluster analysis of periodic signals, proposed the formal conditions which must be satisfied by a period of signal existence, and given some results of analysis of real data recorded in field conditions. Analysis of the results obtained by studying real signals allows us to say that the approach in question can result in close estimations of a locus in time of a pure signal, and in less close estimations of a locus in time of noisy signals.

Our main objective was restricted by development of the method of formalized analysis of periodic signals for estimation of their period of existence. We have not concerned methods of improving signals as it is a theme of separate investigation. We suppose that more exact decisions can be attained by attracting analysis of the left and the right uniformity of cluster families (Znak V. I., 2009) and frequency processing (Znak V. I., 2005). Cluster families, which reflect a locus of a signal on its boundaries, must have a higher uniformity than for others.

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